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- I Source coding and communication system II- Lossless Source Coding : Information Theory III- Lossless Source Coding algorithms
 - Huffman
 - Lempel-Ziv (Welch)
 - Arithmetic Coding









Source coding = data compression To represent the source (data) with the less code symbols as possible and with the highest fidelity (lowest distorsion) as possible

Lossless compression:

Enables error free decoding Unique decodability without ambiguity

Lossy compression:

Distorsion and compression



Claude Elwood Shannon (1916 – 2001),

American electrical engineer and mathematician, has been called "the father of information theory", and was the founder of practical digital circuit design theory.





II- Information Theory

Définition of information quantity

 \rightarrow = doubt quantity, linked to the event probability:

$$i(x) = F(p(x))$$
 with $F(1) = 0$

→ additive quantity:

i(xy)=i(x)+i(y) if x and y independent

Therefore

i(x) = -a.log(p(x))

unity: Binary Unit

Related to the simpliest random experience: the equiprobable binary one

i(pile)=i(face)= -a.log(1/2) = 1 binary unit = 1 binit = 1 bit

By choosing to compute the function « log » in base 2, a=1 !

Thus

$$i(x) = -\log_2(p(x))$$
 bits

Elements of Information Theory, Thomas M. Cover and Joy A.Thomas, John Wiley, 1991.

Other unities exist but « bits »: the most used

C. E. Shannon, « A mathematical theory of communication », *Bell System Technical Journal*, vol. 27, pp. 379-423 and 623-656, July and October, 1948. See on web site <u>http://cm.bell-labs.com/cm/ms/what/shannonday/shannon1948.pdf</u>



of Information Theory

Information

Theory

Shannon

(1916 -2001) The « father »



- It will be crucial to be able to quantify the amount of randomness of a probability distribution (a source !)
- Definition: The **entropy** H(X) of a discrete random variable X is defined by (also denoted H(p)):

$$H(X) = -\sum_{x} p(x) \log_2 p(x)$$

• The entropy of a distribution is expressed in *bits*.

You can view *H* as the expectation of $-\log (p(x))$: H(X) = $-\Sigma_x p(x) \log p(x) = E_p \{ -\log p(X) \}.$

Some Properties of H

- always $H(X) \ge 0$.
- H(X)=0 iif X is a 'deterministic variable' with p(x)=1 for one specific value $x \in \mathcal{X}$.
- If p(x) = 1/D for D different values $x \in \mathcal{X}$, then $H(X) = \log D$.
- $H(X) \le \log(\text{number of } x \in H \text{ with } p(x) > 0);$ **H(X)** *maximum for equiprobability statistics*.
- Decomposition increases entropy H(p1,p2,...,pN) > H(P,Q)

Entropy of a Bit

- •A completely random bit with $p=(\frac{1}{2},\frac{1}{2})$ has H(p) = $-(\frac{1}{2} \log \frac{1}{2} + \frac{1}{2} \log \frac{1}{2}) = -(-\frac{1}{2} + -\frac{1}{2}) = 1$.
- •A deterministic bit with p=(1,0) has H(p) = -(1 log 1 + 0 log 0) = -(0+0) = 0.
- •A biased bit with p=(0.1,0.9) has H(p) = 0.468996...

•In general, the entropy looks as follows as a function of $0 \le \Pr{X=1} \le 1$:





The expected entropy of Y after we have observed a value x∈X, is called the *conditional entropy* H(Y|X)

$$\begin{split} \mathsf{H}(\mathsf{Y}|\mathsf{X}) &= \sum_{\mathsf{x}} \mathsf{p}(\mathsf{x}) \cdot \mathsf{H}(\mathsf{Y}|\mathsf{X} = \mathsf{x}) \\ &= -\sum_{\mathsf{x}} \mathsf{p}(\mathsf{x}) \cdot \sum_{\mathsf{y}} \mathsf{p}(\mathsf{y}|\mathsf{x}) \log \mathsf{p}(\mathsf{y}|\mathsf{x}) \\ &= -\sum_{\mathsf{x},\mathsf{y}} \mathsf{p}(\mathsf{x},\mathsf{y}) \log \mathsf{p}(\mathsf{y}|\mathsf{x}) \\ &= -\mathsf{E}_{\mathsf{p}(\mathsf{x},\mathsf{y})} \log \mathsf{p}(\mathsf{Y}|\mathsf{X}) \end{split}$$
 Also $\mathsf{H}(\mathsf{X}/\mathsf{Y})$

Chain rule: H(X,Y) = H(X)+H(Y|X) = H(Y)+H(X|Y). ⁸

Mutual Information



- For two variables X,Y the mutual information I(X;Y) is the amount of certainty regarding X that we learned after observing Y. Hence I(X;Y) = H(X)–H(X|Y).
- Note that now X and Y can be interchanged using the chain rule: I(X;Y) = H(X) – H(X | Y)

$$= H(X, Y) - H(Y | X) - H(X | Y)$$
$$= H(Y) - H(Y | X)$$
$$= I(Y; X)$$

• Think of I(X;Y) as the 'overlap' between X and Y.





The channel capacity C is the maximum over all possible p(x): C = max_{p(x)} I(X;Y).

[Cover & Thomas, Section 8.3]: $C \ge 0$ and $C \le \log |X|$ and $C \le \log |Y|$ as $I(X,Y) \le \log |X|$, $\log |Y|$.

Some Example Capacities

- A noiseless binary channel has H(X|Y)=0, hence for the mutual information I(X;Y)=H(X), which is maximized by p(0)=p(1)=¹/₂.
 Hence C = max_p I(X;Y) = 1 bit.
- For a noisy, symmetric binary channel we have H(Y|X=x)=H(p), hence I(X;Y) = H(Y)-H(p).

Hence $C = \max_{p} I(X;Y)$ is obtained for H(Y)=1 (again $p(0)=p(1)=\frac{1}{2}$), such that C = 1-H(p).



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III- Coding algorithms

Discrete source alphabet: $X = \{x_1, x_2, ..., x_N\}$ N messages Entropy H(X) (bits) Discrete channel with all

Discrete channel with alphabet: $U = \{u_1, u_2, ..., u_D\}$ Capacity C (bits) D symbols

Often N > D Coding : x_k => codeword : $m_k = u_{n_1}u_{n_2} \dots u_{n_k}$ n_k : length of the codeword



Entropy = minimum mean length of binary code

Efficacity $E = H(X)/n \log_2(D)$ Redundancy $\rho = 1 - E$

CODE PROPERTIES

- A code is non-singular if every element of S_x maps into a different string in D^{*}, i.e., x_i ≠ x_j ⇒ C(x_i) ≠ C(x_j).
- A code is a uniquely decodable code if its extension is non-singular.
- A code is a prefix code or an instantaneous code if no codeword is a prefix of any other codeword.

(no codeword is the beginning of an other)

There exists almost one instantaneous code such that $H(X) / log_2(D) \le \overline{n} \le H(X) / log_2(D) + 1$

INTERNATIONAL MORSE CODE

- 1. A dash is equal to three dots.
- 2. The space between parts of the same letter is equal to one dot.
- 3. The space between two letters is equal to three dots.
- 4. The space between two words is equal to five dots.



 Huffman codes are special prefix codes that can be shown to be optimal (minimize average codeword length)



Huffman Algorithm:

1) Arrange source letters in decreasing order of probability $(p_1 \ge p_2 ... \ge p_k)$

- 2) Assign '0' to the last digit of X_k and '1' to the last digit of X_{k-1}
- 3) Combine pk and pk-1 to form a new set of probabilities

4) If left with just one letter then done, otherwise go to step 1 and repeat

Huffman code: an example



Huffman mean length: 2.61 H(X)=2.55 bits thus an efficiency of E=97.8%

What else ?



Can we use any channel for any source ?





Shannon's noisy channel-coding theorem shows that unreliable channels can be used for reliable communication if we code our messages cleverly.
More specifically, the theorem states that each (discrete, memoryless) channel has a *capacity* C'≥0, such that each "bits per transmission" *rate*

R < C' (bits / s)

is achievable if we use long enough codes.

(Achievable here means that the error probability $\lambda^{(n)}$ tends to zero as the length n of the codes grows.)





Code has to be efficient enough !... And if not ?



Noiseless coding theorem

Source $X = \{x_1, x_2, ..., x_N\}$. It can be shown: There exists a « good » code with mean length n such that $H(X)/log_2(D) \le n \le H(X)/log_2(D) + 1$ Now code the « source extension » $X_k = \{x_1x_1...x_1, ..., x_Nx_N...x_N\}$ Bloc of k messages

 $H(X)/log_2(D) \le n \le H(X)/log_2(D) + 1/k$

Huffman Coding: The Retired Champion

- Replacing an input symbol with a codeword
- Need a probability distribution
- Hard to adapt to changing statistics
- Need to store the codeword table
- Minimum codeword length is 1 bit



Huffman Coding (1952) : optimal code if source statistics known

If unknown, no more optimal ...

Arithmetic Coding: The Rising Star

- Replace the entire input with a single floating-point number
- Adaptive coding is very easy
- Fractional codeword length

Dictionnary-based coding (Ziv-Lempel and &): another alternative

No statistics estimation !



